Emotion Recognition from Speech Signals Using DCNN with Hybrid GA-GWO Algorithm

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Abstract: In recent days, from the speech signal the recognition of emotion is considered as an extensive advanced investigation subject because the speech signal is considered as the rapid and natural method to communicate with humans. Numerous examinations have been progressed related to this topic. This paper develops the emotions recognition from the speech signal in an accurate way, with the knowledge of numerous examined models. Therefore, to study the multimodal fusion of speech features, a Deep Convolutional Neural Network model is proposed. Moreover, the hybrid Genetic Algorithm (GA)-Grey Wolf Optimization (GWO) algorithm is presented that is the combination of both the GA and GWO technique features towards training the network. Finally, the developed recognition model is verified and compared with the existing techniques in correlation with diverse performance measures such as Accuracy, Sensitivity, Precision, Specificity, False Positive Rate (FPR), False Discovery Rate (FDR), False Negative Rate (FNR), F1Score, Negative Predictive Value (NPV), and Matthews correlation coefficient (MCC).

Keywords: Speech; Recognition Model; Emotion; Neural Network; Optimization

Nomenclature

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<td>DAE</td>
<td>Denoising Auto Encoder</td>
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<td>Grey Wolf Optimization</td>
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<td>IBM</td>
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<td>PSD</td>
<td>Power Spectral Density</td>
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<td>LSTM</td>
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<td>FT</td>
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<td>Ideal Ratio Mask</td>
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<td>DFT</td>
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<td>Minimum Mean-Square Error</td>
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<td>STFT</td>
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<td>SNR</td>
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<td>Residual Long Short-Term Memory</td>
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<td>PASTFT</td>
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<td>WF</td>
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<td>DCNN</td>
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<td>PSO</td>
<td>Particle Swarm Optimization</td>
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<td>IR</td>
<td>Impulse Response</td>
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1. Introduction

In several day-to-day life speech communication technologies, speech enhancement is considered as specific attention, whereas noise environments challenges urge for a consistent speech processing device performance [1]. For speech enhancement when important attempts have been committed within the past decades, in order to find innovative ideas to push the restriction of the attainable noise reduction.
For instance, in a cocktail party case, attracts a huge amount of attention with many possible applications in hearing technologies and mobile phones.

By humans or enhance decoding by systems the noisy speech signals processing is to enhance their perception is how speech enhancement deals with. A mode of speech enhancement techniques is to enhance the system performance while the input of the speech is harmed by noise [2]. Usually, it is difficult to keep speech undistorted when minimizing noise and hence, restrictions on the performance of the speech enhancement systems, the compromise among noise reduction and speech distortion [3]. To high SNR, for distorted speech with medium, the objective will be able to create subjectively natural signal by minimizing noise level and for those with minimum SNR [6]. Moreover, when preserving the intelligibility, the purpose of this technique could be to minimize the noise level. The general factor which creates the speech’s quality degradation and intelligibility is environment noise that can be non-stationary or stationary and is understood to be additive and uncorrelated to the speech signal [4]. An extensive categorization of speech enhancement techniques can be specified as temporal processing and spectral processing techniques. In frequency domain, the degraded speech leaves during processing in the spectral processing techniques. However, for temporal processing technique, the processing will be in time domain [5].

The research concentrates on the speech enhancement community, which turns into deep learning techniques. Recently for the speech enhancements, the deep learning techniques have been utilized, which has shown outstanding performance [8]. The NN was exploited for clean speech spectra from noisy speech spectra [18] [19] as non-linear maps. By utilizing clean speech and noisy pairs, a DAE was pretrained for this task [9]. A non-causal NN clean speech spectrum estimator was presented which generated improved speech with maximum objective quality scores [10], that later integrated multi-objective IBM-based post-processing [10]. In order to calculate the time-frequency masks, the neural networks were exploited. Recently, an LSTM network was utilized to calculate the IRM [11]. It is significant to notice that both noise and speech signals are extremely non-stationary, that is, the signal energy distribution over frequency, which is not constant in time. Moreover, the smoothing operations, which does not consider, might produce biased approximates.

To ease this restriction, a supervised learning technique was exploited whereas some hours of the training data were frequently exploited to calculate the association among noisy signals as well as equivalent clean signals [12]. By real data, the signal estimation rules are constructed, also discrepancy among the expected and actual model can be minimized. It tends to enhance speech enhancement performance were adequate training data state is present. Lately, in order to accurately model the clean speech spectrum, a deep autoencoder framework was used [13].

The main objective of this paper is to develop a model for recognize the emotions from the speech signals in an accurate manner. Hence, a DCNN model is presented in order to study the multimodal fusion of speech features. Here, the speech features are prosodic features, cepstral, and NMF. Multimodal features can represent each speech signal at unique margin that is extremely suggested by numerous researchers.

### 2. Literature Review

In 2019, Ki-Seung Lee et al [1], examined the utilization of the ultrasonic doppler frequency shifts, which occurred using movements in facial. It was used for improving the audio speech adulterated using high levels of acoustic noise. At first, the conventional signals were demodulated and it was transformed into a parameter of the spectral feature. From the UDS, the spectral feature was derived, and from noisy speech, it was concatenated with spectral features that were subsequently exploited to estimate the spectrum magnitude of clean speech. In this estimation, a nonlinear regression technique was exploited whereas the connection among audio-UDS features as well as the equivalent clean speech was indicated using DNN.

In 2019, Johannes Stahl and Pejman Mowlaee [2] presented a PASTFT model in order to attain a signal-dependent time-frequency depiction for the input signal. The association of inter-frame for the consecutive speech was analyzed, the DFT bins occurring from the harmonic signal modeling and the PASTFT. If the phase advanced development using the harmonic nature of the speech signal was considered and this analysis shows notable correlation. Therefore, consecutive speech DFT bins represented as composite-valued autoregressive procedures were modeled and to integrate the harmonic stage development into a state-transition designed.

In 2019, Aaron Nicolson and Kuljip K. Paliwal [3] investigated deep learning algorithm for the MMSE method, with the aim of processing comprehensible improved speech with maximum quality. As the speech enhancement performance for an MMSE technique enhances with the precise of the exploited a priori SNR estimator, a ResLSTM network was exploited here to precisely calculate a priori SNR.
MMSE algorithms utilizing the ResLSTM a priori SNR estimator were estimated by objective and subjective measures of intelligibility and speech quality.

In 2017, Qi He, Feng Bao and Changchun Bao [4], proposed a novel wiener filtering speech enhancement technique in order to estimate the short period of time linear predictive parameters for noise and speech in the codebook-driven. A pre-trained spectral form codebook for the speech was exploited in order to model a priori instruction regarding coefficients of linear predictive for speech. In the presented technique, a multiplicative update rule was exploited to calculate the linear predictive gains high precisely were used.

In 2017, Ji Ming and Danny Crookes [5], developed a novel technique that aspires to minimize or efficiently remove requirements. By exploiting the ZNCC, it was demonstrated that was in contrast to the measure, and by expanding the efficient speech segment length corresponding to sentence long speech statement. From the noise, it was probable to attain an accurate speech calculation without necessitating the particular knowledge regarding the noise. A new realization which incorporates the full-sentence speech correlation with uncontaminated recognition of speech, established as a constrained maximization issue, to control the data sparsity issue.

In 2018, Robert Rehr and Timo Gerkmann [6], presented a theoretical and experimental estimate of the MLSE-based methods. The super-Gaussian prior’s permits for a minimization of noise among speech spectral harmonics that was not attainable by Gaussian estimators like the WF. A deep neural network on the basis of the low-rank nonnegative matrix factorization and a phoneme classifier model was exploited for the evaluation as instances of MLSE based techniques.

In 2018, Johannes Stahl and Pejman Mowlaee [7], presented a new outlook producing three major contributions. Initially, a pitch-synchronous signal indication was contemplated and exhibited to be dominant for the calculation of the harmonic parameters model. Subsequently, the harmonic amplitudes were modeled in voiced speech as arbitrary variables with frequency bin determined by Gamma distributions. At last, for the different methods, the distinct estimators of unvoiced speech, voiced speech, as well as speech non-presence was derived.

3 Emotional Speech Signal Analysis

3.1 Analysis of NMF

Let us assume the non-negative data vector of D(n) as D ∈ T=vj;1 × j, which represents the 1-dimensional samples, the main contribution of NMF [14] is to recognize two non-negative matrices that are stated in eq.(1) and (2), Here v indicates the dimensional space.

\[
X \in T=v^v
\]

\[
Y \in T=v^v
\]

By the non-negative constraints, the parts-based indication is achieved as they permit one and only addition and not either combinations or subtractive. It is necessary to describe the function of the cost to recognize the approximate factorization that enumerates the approximation quality. The most well-known cost models are:

(a) The squared Euclidean distance among YZ as well as S is described in eq. (3), here | · | Fb indicates the matrix Frobenius norm.

\[
| | D- XY | | Fb = \sum_{m=1}^{v} | | S_m - (XY)_m | |^2
\]

(b) In eq. (4), the general Kullback-Leibler divergence K between R and XY is described.

\[
K(R | | XY) = \sum_{m=1}^{v} \left( R_m \log \frac{R_m}{(XY)_m} - R_m + (XY)_m \right)
\]

By multiplicative technique, both the functions of cost are resolved [15]. For the square Euclidean distance, the update instruction for non-negative matrices Y_m and X_m are stated in eq. (5) and (6) as well as the updating rule for the divergence of Y_m and X_m are stated in eq. (7) and (8), whereas m = 1...v . In eq. (9), the feature of NMF \( NMF, (\alpha) \) is stated.

\[
Y_m \leftarrow \frac{X^T R_m}{X^T (XY)_m} Y_m
\]
\begin{align*}
X_m & \leftarrow \left( \frac{RY^T}{XY^T} \right) Y_m \\
Y_m & \leftarrow \sum_i \frac{X_i R_{lm}/(XY)_{lm}}{\sum_i Y_i} \quad \text{(6)} \\
Z_m & \leftarrow \sum_i \frac{Z_{ml} R_{ml}/(XY)_{ml}}{\sum_i Y_{ml}} \quad \text{(7)} \\
NM_i(n) & = (X_m, Y_m) \quad \text{(8)} \\
\end{align*}

3.2 Cepstral Analysis

Consider the speech signal as \( S_s(n) \), that is attained from the complexity of two signals \( h(n) \) and \( \hat{j}(n) \) as the summation of two signals as well as it is stated in eq. (10). Here \( \hat{S}_s(n) \) represents the complex cepstrum and \( \hat{S}_s(n) \) is indicated in eq. (17).

\[
S_s(n) = h(n) + j(n) \rightarrow \hat{S}_s(n) = \hat{h}(n) + \hat{j}(n)
\]

In eq. (10), \( h(n) \) represents the speech portion and \( j(n) \) represents the noiseless piece of the speech recording correspondingly and eq. (11) states the analysis of the cepstral, and eq. (12) states the log of the signal \( SL(z) \).

\[
\hat{S}_s(n) = \hat{S}_{s_1}(n) \ast \hat{S}_{s_2}(n) \iff SL(z) = SL_1(z)SL_2(z)
\]

\[
\log |SL(z)| = \log |SL_1(z)| + \log |SL_2(z)| = \hat{SL}(z)
\]

If the transform \( Z \) is reasonable and the composite log is characteristic, subsequently the two convolved signals \( \hat{S}_{s_1}(n) \) and \( \hat{S}_{s_2}(n) \) are added, as well as it is demonstrated in Eq. (13).

\[
\hat{S}_s(n) = \hat{S}_{s_1}(n) + \hat{S}_{s_2}(n)
\]

In eq. (14), the signal \( S_s(n) \) is limited to have poles and zeros into the unit circle, here \( \log |SL(u)| \) is the complex logarithm of \( SL(u) \).

\[
\log |SL(U)| = \log |SL(u)| + 1 < SL(u)
\]

Eq. (15) states if \( SL(u) = SL_1(u)SL_2(u) \) subsequently \( \log |SL(u)| \) is indicated. In eq. (16), the real cepstrum \( RS_s(n) \) is stated where the magnitude of \( RS_s(n) \) is non-negative and real. The \( \hat{S}_s(n) \) complex cepstrum is stated in eq. (17), whereas the stage is indicated as \( \text{arg} \), \( \log |SL(u)| \) and \( |SL(e^{i\theta})| \) are the log spectrum of the signal. This is intricate due to it uses the complex logarithm. Additionally, the compound cepstrum of the real series is real.

\[
\log |SL(u)| = \log |SL_1(u)| + \log |SL_2(u)|\]

\[
RS_s(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \left[ \log |SL(e^{iu})| \right] e^{i\nu} du
\]

\[
\hat{S}_s(n) = \frac{1}{2\pi} \left[ \log |SL(e^{iu})| + \text{arg} |SL(e^{iu})| \right] e^{i\nu} du
\]

In reality, the real cepstrum is the even part of \( \hat{S}_s(n) \), as well as it is shown in eq. (18). In general, speech processing \( RS_s(n) \) is exploited that is attained by using an inverse FT of the log spectrum of the signal.

\[
RS_s(n) = \frac{\hat{S}_s(n) + \hat{S}_s(n)}{2}
\]

Eq. (19) represents the FT is replaced by DFT in digital signals. Here, \( \hat{SL}_a(g) \) is indicated as the sampled version of \( \hat{Y}(e^{kw}) \) and hence \( \hat{S}_s(n) \) is stated as demonstrated in Eq. (20), and \( N \) indicates the period. Likewise, the misidentification of a signal frequency, by repeating the cepstrum with \( N \), the \( RS_s(n) \) cepstral feature of \( S_s(n) \) is stated in Eq. (21).
\[
\text{SL}_t (g) = \sum_{n=0}^{N-1} S_n(n) e^{-\frac{2\pi g n}{N}} \quad 0 \leq g \leq N - 1
\]
\[
\hat{\text{SL}}_t (g) = \log |\text{SL}_t (g)| \quad 0 \leq g \leq N - 1
\]
\[
\hat{S}_n (n) = \frac{1}{N} \sum_{g=0}^{N-1} \hat{S}_i(g) e^{-\frac{2\pi g n}{N}} \quad 0 \leq n \leq N - 1
\]
\[
\hat{S}_n (n) = \sum_{p=-\infty}^{\infty} \hat{S}_n (n + mN)
\]
\[
\text{RS}_t (n) = \frac{1}{N} \sum_{g=0}^{N-1} \log |\text{SL}_t (g)| e^{-\frac{2\pi g n}{N}} \quad 0 \leq n \leq N - 1
\]

### 3.3 Analysis of Pitch

The entire estimation of pitch filter consists of some steps, and it is stated as below:

(a) At first to transform the \( t_r (n) \) to the time-frequency domain with the help of STFT) \( = TF_r (f) \) that is stated in eq. (22). Here, \( V_r (f) \) indicates the PSD of the exasperating noise as well as \( t_r (n) \), the power of the \( P^{th} \) harmonic, \( f_{r0} \) as well as \( n \) indicates the time and frequency for the ideal periodic source.

\[
\text{TF}_r (f) = \sum_{p=1}^{P} t_r (n) \Delta (f - p f_{r0}) + \text{IR} (f)
\]  

(b) In this process, the PSD of each frame against a Log spaced frequency grid \( \text{TF}_r (l) \) that is stated in eq. (23), here \( l = \log f \). Moreover, the harmonic spacing is unconventional to \( f_{r0} \) using convolving \( \text{TF}_r (f) \) with IR \( f \) the energy should be integrated that is stated in eq. (24).

\[
\text{TF}_r (l) = \sum_{p=1}^{P} t_r (n) \Delta (l - \log p - \log f_{r0}) + V_r (l)
\]
\[
I(l) = \sum_{p=1}^{P} \Delta (l - \log p)
\]

(c) As stated in eq. (25), formulate \( \beta_r (l) \). To decide the compression exponent \( \beta_r (l) \), in both the log-frequency and time \( L_P (l) \), the first procedure is to construct the smoothened spectrum \( \overline{F}_r (l) \) using low pass filtering \( F_r (l) \). Both \( F_r (l) \) and \( \overline{F}_r (l) \) are normalized to the power of \( L_P (l) \) as well as sets the compression exponent \( \beta_r (l) \). So, the normalized smoothened spectrum \( \overline{F}_r (l) \) equals \( L_P (l) \) and the compressed PSD formulation \( F_r (l) \) is performed as stated in eq. (26)

\[
\beta_r (l) = \frac{\log L_P (l)}{\log F_r (l)}
\]
\[
F_r (l) = F_r (l^\beta_r (l))
\]

(d) Here, \( F_r (l) \) is convolved with \( I(1) \) and selects the maximum peak in the sensible range as \( P_r (n) \), the assessed pitch.

\[
\text{RR}_r (n) = |F_r (l) \ast I(l)|
\]

The initial process is feature extraction, in that from the input \( t_r (n) \), the features are extracted. The resultant features such as NMF feature \( \text{NMF}_r (n) \), Cepstral feature \( \text{RS}_r (n) \), and pitch feature \( \text{RR}_r (n) \) are stated as the input to the dimensionality minimized phase. The produced dimensional minimized features are integrated and stated as the input to the classification procedure that outcome the classified output. Fig 1 demonstrates the architectural diagram of the proposed model.
4. Detailed Analysis of the Proposed Algorithm

4.1 Feature Extraction

The proposed method consists of three features such as RS\(_t\)\(_n\), NM\(_t\)\(_n\) and RR\(_t\)\(_n\) are extracted that is demonstrated in Eq. (9), (21) and (27). The feature extraction is stated to the subsequent procedure in order to minimize the dimensions of the features.

4.2 Reduction of Dimensionality

In this phase, the extracted CF\(_t\)\(_n\), NM\(_t\)\(_n\) and RR\(_t\)\(_n\) are given as the input. Here, the well-known process called PCA is exploited to reduce the dimensions of the features. A collection of \(dv\) data vectors \(g_1, g_2, \ldots, g_{dv}\) are selected from the extracted feature and \(g_u\) indicates the well-defined group analysis of variables \(u\). Subsequently, the empirical mean formulation is performed with all column cl = 1, \ldots, \(u\), and the ensuing mean value is positioned in the \(e_u\) with \(u \times 1\) dimensions, that is stated in eq. (28), the computation of mean deviation is done as per eq. (29). Here, \(F\) indicates the \(d \times u\) matrix and \(k\) represents the column vector \(d \times 1\) of all 1s: \(k[i] = 1\).

\[
e_u[cl] = \frac{1}{d_v} \sum_{i=1}^{d_v} RS[i, cl]
\]

\[
F = S - ke_k^T
\]

As demonstrated in eq. (30), the covariance matrix \(CM\) is stated. Moreover, both the formulation of eigenvalues and eigenvector is processed using valuing the \(M\) matrix, that diagonalizes \(CM\) and the formulation is stated in eq. (31)
\[ CM = \frac{1}{d_v - 1} F^T F \]  
\[ P^{-1}CM(P) = D^{(p)} \]  

In eq. (31), \( D^{(p)} \) indicates the diagonal matrix of eigenvalues of \( CM \). In eigenvector matrix \( N \) the column sorting is performed, the eigenvalue matrix minimization of \( D^{(p)} \) is attained. The computation of cumulative energy content is expressed in eq. (32) that embraces the sum of the energy content of eigenvalues from 1 through \( cl \).

\[ ce[cl] = \sum_{d_v = 1}^{cl} D^{(p)}[d_v, d_v], \text{ for } cl = 1, \ldots, u \]  

As \( B \) matrix by storing the \( cl \) column of \( P \), the eigenvectors subset is chosen. To select the cost of \( cl \), the \( l \) vector is exploited. The column of \( \hat{P} \) the matrix \( \hat{P} = Q.B = KKL[R] \) is the vector, as well as in the row matrix \( R \), it decides the Kosambi-Karhunen-Loeve transform. In eq. (33), the \( Q \) matrix is stated, here \( t = t(c) = \sqrt{\frac{1}{\langle CM \rangle[cl, cl]}}, cl = 1, \ldots, v \) as well as this is the generated dimensionally minimized speech signal \( S_{cl}(n) \).

\[ Q = \frac{F}{\sqrt{\sum_{j=t}^{T}}} \]  

For all the extracted features individually the PCA process is exploited, as well as the dimensional minimized signal \( S_{cl}(n) \) is stated as stated in eq. (34), here \( CF^{dv}_{cl}(n), NM^{dv}_{cl}(n), m^{dv}_{cl}(n) \) indicates the dimensional minimized NMF, cepstral, and pitch features in the proposed method.

\[ S^{dv}_{cl}(n) = [CF^{dv}_{cl}(n), NM^{dv}_{cl}(n), m^{dv}_{cl}(n)] \]  

4.3 Emotion Recognition using the Proposed Method

(a) DCNN: In this paper, the deep neural network models, the DCNN is exploited. In recent years, it has been acquired huge popularity in classification. An archetypal DCNN comprises 5 convolutional layers, like three fully-connected layers and three pooling layers [16]. Using competent alternating and pooling layers and stacking convolutional layers, feature learning is attained. The one-dimensional feature vector obtained from plotting two-dimension feature vectors using the polling layers as well as the fully-connected layers follow the convolution layers.

The mapping of output feature maps for the \( m^{th} \) layer \( y^{m} \) can be computed on the basis of the Eq. (35) if there are \( P \) feature maps as the input and \( Q \) filters for convolutional layers.

\[ y^{m}_{j} = f\left( \sum_{i=1}^{P} y^{m-1}_{ij} * r^{m}_{ij} + a^{m}_{ij} \right), j = 1, \ldots, Q \]  

In eq. (35), \( r^{m}_{ij} \) indicates the kernel of the \( j^{th} \) filter associated with the \( i^{th} \) input map, \( y^{m-1}_{ij} \) indicates the \( i^{th} \) input map, \( a^{m}_{ij} \) is the bias equivalent to \( j^{th} \) filter, \( * \) indicates the convolutional operation, and \( f(\ ) \) indicates the activation function. Hence, \( Q \) feature maps are attained as the output. Using eq. (36), the number of all the parameters of a convolutional layer is computed.

\[ PR = N \times (k \times k \times P + 1) \]  

A convolutional layer is gone after by the pooling layer directly holds the activations within the little spatial region. The nature of the operations for pooling is characterized into two types: average and maximum pooling whereas the maximum-pooling unit calculate the utmost for a local patch of unites in a feature map, as well as the average pooling calculates average. The computation procedure of output feature maps of the \( m^{th} \) layer is same as with eq. (35) in a pooling layer.

\[ y^{m}_{j} = f\left( a^{m}_{ij} \downarrow \left| y^{m-1}_{ij} \right| \right), j = 1, \ldots, Q \]  

In eq. (37), \( \downarrow \left( \right) \) indicates the sub-sampling function and \( a^{m}_{ij} \) represents the multiplicative bias equivalent to the \( j^{th} \) filter. Subsequent to the pooling layers and the convolutional layers; from the raw data, the fully connected layer is exploited to classify the features extracted. Into the one-dimensional vector, the learned feature vectors are demolished that is an input of the fully connected layers. For the input vector each value the is linked for every value of the output vector by one neuron in a fully
connected layer. If the lengths of the output and input vectors are $Q$ and $P$ correspondingly; as eq. (38) the output vector of the $m^{th}$ layer is computed.

$$y^m_j = f \sum_{i=1}^{P} y^{m-1}_i \times z^m_i + a^m_j, \ j = 1, \ldots, Q$$  \hspace{1cm} (38)$$

In eq. (38) $z^m_i$ indicates the weight of the $j^{th}$ output value linked with the $i^{th}$ input value. The calculation for the number of the entire parameters of a fully connected layer is stated in eq. (39).

$$M = P \times Q + 1$$  \hspace{1cm} (39)$$

(b) Hybrid GWO-GA:

In this paper, the GWO is integrated GA to enhance the effectiveness of the method [17]. Assume that the present population as $P = (Y_1, Y_2, \ldots, Y_D)$, the fitness of the individual $Y_i$ is $f_i$. Compute the fitness $f_i (i = 1, 2, \ldots, D)$ of each individual and organize them in descending order. Choose the individuals with the maximum fitness to copy directly into the subsequent generation of the population. Compute the total fitness $T$ of the residual individuals and the probability $P_{pi}$ that each individual is chosen.

$$T = \sum_{i=1}^{D-1} f_i (i = 1, 2, \ldots, D - 1)$$  \hspace{1cm} (40)$$

$$P_{pi} = \frac{f_i}{\sum_{i=1}^{D-1} f_i} (i = 1, 2, \ldots, D - 1)$$  \hspace{1cm} (41)$$

For each individual, compute the cumulative fitness value $c_i$, and subsequently, the selection operation is done in the way of the stake roulette until the number of individuals in the children population is reliable with the parent population.

$$c_i = \sum_{i=1}^{D-1} \frac{f_i}{T} (i = 1, 2, \ldots, D - 1)$$  \hspace{1cm} (42)$$

In the subpopulation, each individual is cross-operated in a linear crossover way in the proposed method. For each individual $y_i$ in the subpopulation, generate a corresponding arbitrary number $r_i \in (0, 1)$. While the random number $r_i$ is lesser than the $C_p$ crossover probability, the equivalent individual $y_i$ is paired for cross-operation. Here, the crossover operators are $(c_1, c_2)$. Here, generate a random number $\eta \in (0, 1)$ and it consists of two children $cl^1[cl^1_1, \ldots, cl^1_D]$, $cl^2[cl^2_1, \ldots, cl^2_D]$, which are produced by two parents $pr^1[pr^1_1, \ldots, pr^1_D]$ and $pr^2[pr^2_1, \ldots, pr^2_D]$.

$$cl^1_i = \eta pr^1_i + (1 - \eta)pr^2_i, \ i = 1, 2, \ldots, D$$  \hspace{1cm} (43)$$

$$cl^2_i = \eta pr^2_i + (1 - \eta)pr^1_i, \ i = 1, 2, \ldots, D$$  \hspace{1cm} (44)$$

The best individual is $y_i (y_1, y_2, \ldots, y_D)$ with the mutation probability $M_p$, the mutation operation is done on $y_i$ namely, choose a gene $y_m$ from the best individual with probability $M_p$, in place of the gene $y_m$ with an arbitrary number among lower and upper bounds to create a new individual $y'_i = (y'_1, y'_2, \ldots, y'_D)$. Eq. (45), the precise operation is described.

$$y'_i = \begin{cases} y_i & \text{if } i = m \\ y_i + \eta \ast (v - 1) & \text{if } i \neq m \end{cases}$$  \hspace{1cm} (45)$$

In eq. (45), $\eta$ represents a random number in $[0, 1]$, and $v$ and $l$ represents the upper and lower bounds of the individual $y_i$, correspondingly.

5. Results and Discussions

5.1 Experimental Procedure

In this paper, two databases are exploited such as benchmark and Marathi database. Here, both the databases are obtained from http://neuron.arts.ryerson.ca/ravdess/index.php. The benchmark database comprises of numerous emotions such as 96 happy, angry, surprise, fear, neutral, and sad emotions. Likewise, the Marathi database includes 40 neutral, surprise, angry, sad, fear, and happy emotions.
5.2 Analysis of the Proposed Model

This section exhibits the analysis of the proposed Hybrid GWO-GA technique and conventional techniques such as GWO-NN, GA-NN and PSO-NN models for both the Marathi and benchmark databases. In fig 2, the performance analysis of the proposed technique over the existing techniques for the benchmark databases is shown. Here, the result analysis demonstrates the accuracy of the proposed approach 15% is better than the GWO-NN, 22% better than the GA-NN, and 28% better than the PSO-NN methods. Fig 3 shows the performance analysis of the proposed approach over the existing techniques for the Marathi databases. Here, the result analysis demonstrates the accuracy of the proposed approach 23% is better than the GWO-NN, 27% better than the GA-NN, and 31% better than the PSO-NN methods. The analysis outcome reveals that the proposed technique is better than the existing techniques.

![Fig. 2. Analysis of the proposed method for benchmark database](image1)

![Fig. 3. Analysis of the proposed method for Marathi database](image2)

In both fig 4 and 5, the analysis of the proposed feature techniques such as an amalgamation of NMF, cepstra, and Pitch over the performance of NMF feature individually, cepstra feature individually and Pitch feature individually for both the Marathi databases and benchmark database. In fig 4, the analysis of the proposed feature technique with the traditional models for the benchmark databases is demonstrated. Here, the result analysis showed that the superiority of the proposed technique over the existing techniques. Here, the result analysis demonstrates the accuracy of the proposed approach 14% is better than the GWO-NN, 16% better than the GA-NN, and 19% better than the PSO-NN methods. Fig 5 demonstrates the performance analysis of the proposed method with the traditional techniques for the Marathi databases. The analysis outcome reveals that the proposed technique is better than traditional techniques. Here, the result analysis demonstrates the accuracy of the proposed approach 31% is better than the GWO-NN, 35% better than the GA-NN, and 37% better than the PSO-NN methods.
6. Conclusion

Generally, speech analysis has become an important element in minimizing the gap among physical and digital world with the increase in man to machine communication. A significant subpart within this domain is the emotion recognition in speech signals that was conventionally examined in linguistics and psychology. Moreover, speech emotion recognition is a field having diverse applications. The main objective of this paper is to present a novel emotion recognition model. Hence, a DCNN model was developed to study the multimodal fusion of speech features. The features such as cepstral, NMF, and prosodic features. Moreover, a hybrid GA-GWO technique was also attained to train the classifier. The proposed recognition method performance was evaluated and compared it with the existing techniques. The proposed method has shown accurate emotions recognition from the speech signal.

References


